# **THEORY OF OPERATION**

**Th[e KT-EX9-2](http://www.analog.com/ADIS16488A?doc=ADIS16488A.pdf) is an autonomous sensor system that starts when it has a valid power supply. After running through its initialization process, it begins sampling, processing, and loading calibrated sensor data into the output registers, which are accessible using the SPI port. The SPI port typically connects to a compatible port on an ARM, using the connections as shown in [Figure 23.](#page-0-0) The four SPI signals facilitate synchronous, serial data communication. Connect the reset line (RST) to VDD or do not connect it to anything for normal operation. The factory default configuration provides users with a data ready signal on the DIO2 pin, which pulses high when new data is available in the output data registers.**



**Figure 23. Electrical Connection Diagram**

<span id="page-0-0"></span>



**Embedded processors typically use control registers to configure their serial ports for communicating with SPI slave devices, such as the KT-EX9-2. [Table 8](#page-0-1) provides a list of settings describing the SPI protocol of the KT-EX9- 2. The initialization routine of the master processor typically establishes these settings using firmware commands to write them into its serial control registers.**



<span id="page-0-1"></span>

### **REGISTER STRUCTURE**

**The register structure and SPI port provide a bridge between the sensor processing system and an external, master processor. It contains both output data and control registers. The output data registers include the latest sensor data, a real-time clock, error flags, alarm flags, and identification data. The control registers include sample rate, filtering, input/output, alarms, calibration, and diagnostic configuration options. All communication between the KT-EX9-2 and an external processor involves either reading or writing to one of the user registers.**



**Figure 24. Basic Operation**

**The register structure uses a paged addressing scheme that is composed of 13 pages, with each page containing 64 register locations. Each register is 16 bits wide, with each byte having its own unique address within the memory map of that page. The SPI port has access to one page at a time, using the bit sequence shown in Figure 25. Select the page to activate for SPI access by writing its code to the PAGE\_ID register. Read the PAGE\_ID register to** 

**determine which page is currently active. [Table 9](#page-1-0) displays the PAGE\_ID contents for each page, together with their basic functions. The PAGE\_ID register is located at Address 0x00 on every page.**



<span id="page-1-0"></span>

Page	<b>PAGE ID</b>	<b>Function</b>			
0	0x00	Output data, clock, identification			
3	0x03	Control: sample rate, filtering, input/output, alarms			
4	0x04	Serial number			
$5 - 12$	$0x05 - 0x0C$	Filter			

**Table 9. User Register Page Assignments**

### **SPI COMMUNICATION**

**If the previous command was a read request, the SPI port supports full duplex communication, which enables external processors to write to DIN while reading DOUT (see [Figure](#page-1-1)  [25\)](#page-1-1). [Figure 25](#page-1-1) provides a guideline for the bit coding on both DIN and DOUT.**





## <span id="page-1-1"></span>**DEVICE CONFIGURATION**

**The SPI provides write access to the control registers, one byte at a time, using the bit assignments shown in Figure 25. Each register has 16 bits, where Bits[7:0] represent the lower address (listed in [Table 10\)](#page-2-0) and Bits[15:8] represent the upper address. Write to the lower byte of a register first, followed by a write to its upper byte (the only register that changes with a single write to its lower byte is the PAGE\_ID register).**

**For a write command, the first bit in the DIN sequence is set to 1. Address Bits[A6:A0] represent the target address, and Data Command Bits[DC7:DC0] represent the data being written to the location[. Figure 26](#page-1-2) provides an example of writing 0x03 to Address 0x00 (PAGE\_ID [7:0]) using DIN = 0x8003. This write command activates the control page for SPI access.**



**Figure 26. SPI Sequence for Activating the Control Page** 

### <span id="page-1-2"></span>**FLASH MEMORY**

**For a flash memory update, ensure that the power supply is** 

**within specification for the entire processing time (see Table 1).**

### **READING SENSOR DATA**

**The KT-EX9-2 automatically starts up and activates Page 0 for data register access. Write 0x00 to the PAGE\_ID register (DIN = 0x8000) to activate Page 0 for data access after accessing any other page.**

**A single register read requires two 16-bit SPI cycles. The first cycle requests the contents of a register using the bit assignments in [Figure 25,](#page-1-1) and then the register contents follow DOUT during the second sequence.**

**The first bit in a DIN command is zero, followed by either the upper or lower address for the register. The last eight bits are don't care, but the SPI requires the full set of 16 SCLKs to receive the request.**

**[Figure 27](#page-1-3) includes two register reads in succession, which starts with DIN = 0x1A00, to request the contents of the Z\_GYRO\_OUT register, and follows with 0x1800, to request the contents of the Z\_GYRO\_LOW register.**



**Figure 27. SPI Read Example**

<span id="page-1-3"></span>**[Figure 28](#page-1-4) provides an example of the four SPI signals when reading PROD\_ID in a repeating pattern. This is an effective pattern to use for troubleshooting the SPI interface setup and communications because the contents of PROD\_ID are predefined and stable.**



<span id="page-1-4"></span>**Figure 28. SPI Read Example, Second 16-Bit Sequence**

# **USER REGISTERS**

<span id="page-2-0"></span>

**Table 10. User Register Memory Map (N/A = Not Applicable)**



<sup>1</sup> R is read only, W is write only, R/W is read and write, and N/A means not applicable.

# **OUTPUT DATA REGISTERS**

**After the KT-EX9-2 completes its start-up process, the PAGE\_ID register contains 0x0000, which sets Page 0 as the active page for SPI access. Page 0 contains the output data, real- time clock, status, and product identification registers.**

### **INERTIAL SENSOR DATA FORMAT**

**The gyroscope, accelerometer, delta angle, delta velocity, and barometer output data registers use a 32-bit, twos complement format. Each output uses two registers to support this resolution. Figure 18 provides an example of how each register contributes to each inertial measurement. In this case, X\_GYRO\_OUT is the most significant word (upper 16 bits), and X\_GYRO\_LOW is the least significant word (lower 16 bits). In many cases, using the most significant word registers alone provides sufficient resolution for preserving key performance metrics.**



<span id="page-4-8"></span>**Figure 29. Gyroscope Output Format Example, DEC\_RATE > 0 The arrows in [Figure 30](#page-4-6) represent the direction of the motion, which produces a positive output response in the output register of each sensor. The accelerometers respond to both dynamic and static forces associated with acceleration, including gravity. When lying perfectly flat, as shown in Figure 19, the z-axis accelerometer output is 1 g, and the x and y accelerometers are 0 g.**

# **ROTATION RATE (GYROSCOPE)**

**The registers that use the x\_GYRO\_OUT format are the primary registers for the gyroscope measurements (see [Table 11](#page-4-1)[,Table 12](#page-4-3) [,Table 13\)](#page-4-5). When processing data from these registers, use a 16-bit, twos complement data format. [Table 14](#page-4-7) provides x\_GYRO\_OUT digital coding examples.**

**Table 11. X\_GYRO\_OUT (Page 0, Base Address = 0x12)**

<span id="page-4-1"></span>

<span id="page-4-7"></span><span id="page-4-5"></span><span id="page-4-3"></span>

**The registers that use the x\_GYRO\_LOW naming format provide additional resolution for the gyroscope measurements (see [Table 15,](#page-4-0)[Table 16](#page-4-2)[,Table 17\)](#page-4-4). The MSB has a weight of 0.01°/sec, and each subsequent bit has ½ the weight of the previous one.**

**Table 15. X\_GYRO\_LOW (Page 0, Base Address = 0x10)**

<span id="page-4-2"></span><span id="page-4-0"></span>

<b>Bits</b>	<b>Description</b>			
[15:0]	X-axis gyroscope data; additional resolution bits			
Table 16. Y_GYRO_LOW (Page 0, Base Address = 0x14)				
<b>Bits</b>	<b>Description</b>			
$[15:0]$	Y-axis gyroscope data; additional resolution bits			
Table 17. Z GYRO LOW (Page 0, Base Address = 0x18)				
<b>Bits</b>	<b>Description</b>			
[15:0]	Z-axis gyroscope data; additional resolution bits			

<span id="page-4-4"></span>

<span id="page-4-6"></span>**Figure 30. Inertial Sensor Direction Reference Diagram**

# **Data Sheet KT-EX9-2**

# **ACCELERATION**

**The registers that use the x\_ACCL\_OUT format are the primary registers for the accelerometer measurements (see [Table 18](#page-5-1)[,Table 19,](#page-5-3)[Table 20\)](#page-5-5). When processing data from these registers, use a 16-bit, twos complement data format. [Table 21](#page-5-9) provides x\_ACCL\_OUT digital coding examples.**

**Table 18. X\_ACCL\_OUT (Page 0, Base Address = 0x1E)**

<span id="page-5-3"></span><span id="page-5-1"></span>

<b>Description</b>				
X-axis accelerometer data; twos complement, $\pm$ 18 g range, 0 g = 0x0000, 1 LSB = 0.8 mg				
Table 19. Y_ACCL_OUT (Page 0, Base Address = 0x22)				
<b>Description</b>				
Y-axis accelerometer data; twos complement, $\pm$ 18 g range, 0 g = 0x0000, 1 LSB = 0.8 mg				
Table 20. Z_ACCL_OUT (Page 0, Base Address = 0x26)				
<b>Description</b>				
Z-axis accelerometer data; twos complement, $\pm$ 18 g range, 0 g = 0x0000, 1 LSB = 0.8 mg				

**Table 21. x\_ACCL\_OUT Data Format Examples**

<span id="page-5-9"></span><span id="page-5-5"></span>

**The registers that use the x\_ACCL\_LOW naming format provide additional resolution for the accelerometer measurements (see [Table 22,](#page-5-0)[Table 23](#page-5-2)[,Table 24\)](#page-5-4). The MSB has a weight of 0.4 mg, and each subsequent bit has ½ the weight of the previous one.**

<span id="page-5-2"></span><span id="page-5-0"></span>

### <span id="page-5-4"></span>**DELTA ANGLES**

**The x\_DELTANG\_OUT registers are the primary output registers for the delta angle calculations. When processing data from these registers, use a 16-bit, twos complement data format (see [Table 25](#page-5-6)[,Table 26,](#page-5-7)[Table 27\)](#page-5-8). [Table 28](#page-5-10)**

**provides x\_DELTANG\_OUT digital coding examples. The delta angle outputs represent an integration of the gyroscope measurements and use the following formula for all three axes (x-axis displayed):**

$$
\Delta \theta_{x,nD} = \frac{1}{2f_s} \times \sum_{d=0}^{D-1} \left( \omega_{x,nD+d} + \omega_{x,nD+d-1} \right)
$$

**where:**

**is the decimation rate = DEC\_RATE + 1.**

 $f_s$  is the sample rate.

**is the incremental variable in the summation formula.**

**is the x-axis rate of rotation (gyroscope).**

**is the sample time, prior to the decimation filter.**

**When using the internal sample clock, is equal to 2460 SPS. When using the external clock option, is equal to the frequency of the external clock, which is limited to a minimum of 2 kHz, to prevent overflow in the x\_DELTANG\_xxx registers at high rotation rates. See [Table 50](#page-8-1) and [Figure 31](#page-9-3) for more information on the DEC\_RATE register (decimation filter).**

**The x\_DELTANG\_LOW registers (see [Table 29,](#page-6-0)[Table 30](#page-6-1)[,Table](#page-6-2)  [31\)](#page-6-2) provide additional resolution bits for the delta angle and combine with the x\_DELTANG\_OUT registers to provide a 32-bit, twos complement number. The MSB in the x\_DELTANG\_LOW registers have a weight of ~0.011° (720/216), and each subsequent bit carries a weight of**  $\frac{1}{2}$  **of the previous one.**

**Table 25. X\_DELTANG\_OUT (Page 0, Base Address = 0x42)**

<span id="page-5-10"></span><span id="page-5-8"></span><span id="page-5-7"></span><span id="page-5-6"></span>

<span id="page-6-0"></span>

### <span id="page-6-2"></span><span id="page-6-1"></span>**DELTA VELOCITY**

**The registers that use the x\_DELTVEL\_OUT format are the primary registers for the delta velocity calculations. When processing data from these registers, use a 16-bit, twos complement data format (see [Table 32,](#page-6-4)[Table 33](#page-6-6)[,Table 34\)](#page-6-8). [Table 35](#page-6-9) provides x\_DELTVEL\_OUT digital coding examples. The delta velocity outputs represent an integration of the accelerometer measurements and use the following formula for all three axes (x-axis displayed):**

$$
\Delta V_{x,nD} = \frac{1}{2 f_x} \times \sum_{d=0}^{D-1} \left( a_{x,nD+d} + a_{x,nD+d-1} \right)
$$

**where:**

**is the decimation rate = DEC\_RATE + 1.**

 $\boldsymbol{f}_s$  is the sample rate.

**is the incremental variable in the summation formula.**

 $a<sub>x</sub>$  is the x-axis linear acceleration.

**is the sample time, prior to the decimation filter.**

**When using the internal sample clock, is equal to 2460 SPS. When using the external clock option, is equal to the frequency of the external clock, which is limited to a minimum of 2 kHz, to prevent overflow in the** 

**x\_DELTVEL\_xxx registers at high rotation rates. See [Table](#page-8-1)  [50](#page-8-1) and [Figure 31](#page-9-3) for more information on the DEC\_RATE register (decimation filter).**

**Table 32. X\_DELTVEL\_OUT (Page 0, Base Address = 0x4E)**

<span id="page-6-4"></span>

<span id="page-6-8"></span><span id="page-6-6"></span>

<span id="page-6-9"></span>**The x\_DELTVEL\_LOW registers (see [Table 36](#page-6-3)[,Table 37,](#page-6-5)[Table](#page-6-7)  [38\)](#page-6-7) provide additional resolution bits for the delta velocity and combine with the x\_DELTVEL\_OUT registers to provide a 32-bit, twos complement number. The MSB in the x\_DELTVEL\_LOW registers have a weight of ~3.052 mm/sec (200 m/sec ÷ 216), and each subsequent bit carries a weight of ½ of the previous one.**

<span id="page-6-5"></span><span id="page-6-3"></span>

### <span id="page-6-7"></span>**MAGNETOMETERS**

**The registers that use the x\_MAGN\_OUT format are the primary registers for the magnetometer measurements. When processing data from these registers, use a 16-bit, twos complement data format. [Table](#page-7-2)** 39**[,Table 40,](#page-7-3)[Table 41](#page-7-4) provide the numerical format for each register, and [Table 42](#page-7-7) provides x\_MAGN\_OUT digital coding examples.**

<span id="page-7-3"></span><span id="page-7-2"></span>

<span id="page-7-7"></span><span id="page-7-4"></span>

### **BAROMETER**

**The BAROM\_OUT register (see [Table 43\)](#page-7-6) and BAROM\_LOW register (see [Table 45\)](#page-7-5) provide access to the barometric pressure data. These two registers combine to provide a 32 bit, twos complement format. Some applications can use BAROM\_OUT by itself. For cases where the finer resolution available from BAROM\_LOW is valuable, combine them in the same manner as the gyroscopes (see [Figure 29\)](#page-4-8)[. Table](#page-7-6)  [43](#page-7-6) provides the numerical format for BAROM\_ OUT, and [Table 44](#page-7-8) provides digital coding examples.**

**Table 43. BAROM\_OUT (Page 0, Base Address = 0x30)**

<span id="page-7-8"></span><span id="page-7-6"></span>



**The BAROM\_LOW register provides additional resolution for the barometric pressure measurement. The MSB has a weight of 20 μbar, and each subsequent bit carries a weight of ½ of the previous one.**

<span id="page-7-5"></span>

# **INTERNAL TEMPERATURE**

**The TEMP\_OUT register provides an internal temperature measurement for observing relative temperature changes inside the KT-EX9-2 (see [Table 46\)](#page-7-1). [Table 47](#page-7-9) provides TEMP\_OUT digital coding examples. Note that this temperature reflects a higher temperature than that of ambient temperature, due to self heating.**

<span id="page-7-9"></span><span id="page-7-1"></span>

**The ALM\_STS register in [Table 48p](#page-7-0)rovides the alarm bits for the programmable alarm levels of each sensor. Note that reading the ALM\_STS register causes all of its bits to restore to 0. The bits only return to 1 if the error condition persists.**

**Table 48. ALM\_STS (Page 0, Base Address = 0x0C)**

<span id="page-7-0"></span>

# **PRODUCT IDENTIFICATION**

**The PROD\_ID register (see [Table 49\)](#page-8-0) contains the binary equivalent of the device number (16,488 = 0x4068).**

**Table 49. PROD\_ID (Page 0, Base Address = 0x7E)**

<span id="page-8-0"></span>

# **DIGITAL SIGNAL PROCESSING**

### **GYROSCOPES/ACCELEROMETERS**

**[Figure 31](#page-9-3) provides a signal flow diagram for all of the components and settings that influence the frequency response for the accelerometers and gyroscopes. The sample rate for each accelerometer and gyroscope is 2.46 kHz. Each sensor has its own averaging/decimation filter stage. When using the external clock option (FNCTIO\_CTRL[7:4], see [Table 67\)](#page-12-0), the input clock drives a sample rate of 2.46 kSPS.**

### **AVERAGING/DECIMATION FILTER**

**The DEC\_RATE register (see [Table 50\)](#page-8-1) provides user control for the final filter stage (see [Figure 31\)](#page-9-3), which averages and decimates the accelerometers, gyroscopes, delta angle, and delta velocity data. The output sample rate is equal to 2460/(DEC\_RATE + 1).**

**When using the external clock option (FNCTIO\_CTRL[7:4], se[e Table 67\)](#page-12-0)), replace the 2460 number in this relationship with the input clock frequency. For example, turn to Page 3 (DIN = 0x8003), and set DEC\_RATE = 0x18 (DIN = 0x8C18, then DIN = 0x8D00) to reduce the output sample rate to 98.4 SPS (2460 ÷ 25).**

**Table 50. DEC\_RATE (Page 3, Base Address = 0x0C)**

<span id="page-8-1"></span>

### **MAGNETOMETER/BAROMETER**

**The update rates for the magnetometer and barometers do not change with the DEC\_RATE register settings. The magnetometer and barometer sampling frequency is 50Hz. It is not configurable via the SEQ\_CNT register.**





## <span id="page-9-3"></span>**FIR FILTER BANKS**

**The KT-EX9-2 provides four configurable, 120-tap FIR filter banks. Each coefficient is 16 bits wide and occupies its own register location for each page. When designing a FIR filter for these banks, use a sample rate of 2.46 kHz and scale the coefficients so that their sum equals 32,768. For filter designs that have less than 120 taps, load the coefficients into the lower portion of the filter and start with Coefficient 1. To prevent adding phase delay to the response, ensure that all unused taps are equal to zero.**

**The FILTR\_BNK\_x registers provide three bits per sensor, which configure the filter bank (A, B, C, D) and turn filtering on and off. For example, turn to Page 3 (DIN = 0x8003), then write 0x0057 to FILTR\_BNK\_0 (DIN = 0x9657, DIN = 0x9700) to set the x-axis gyroscope to use the FIR filter in Bank D, to set the y-axis gyroscope to use the FIR filter in Bank B, and to enable these FIR filters in both x- and y-axis gyroscopes. Note that the filter settings update after writing to the upper byte; therefore, always configure the lower byte first. In cases that require configuration to only the lower byte of either FILTR\_BNK\_0 or FILTR\_BNK\_1, complete the process by writing 0x00 to the upper byte.**

**Table 51. FILTR\_BNK\_0 (Page 3, Base Address = 0x16)**

<span id="page-9-0"></span>

<span id="page-9-1"></span>

### **FILTER MEMORY ORGANIZATION**

**Each filter bank uses two pages of the user register structure. See [Table 53,](#page-9-2)[Table 54](#page-10-0)[,Table 55](#page-10-1) an[d Table 56](#page-10-2) for the register addresses in each filter bank.**

<span id="page-9-2"></span>

Page	<b>PAGE ID</b>	<b>Address</b>	Register		
5	0x05	0x00	PAGE ID		
5	0x05	0x02 to 0x07	Not used		
5	0x05	0x08	FIR COEF A000		
5	0x05	0x0A	FIR COEF A001		
5	0x05	$0x0C$ to $0x7C$	FIR COEF A002 to		
			FIR COEF A058		
5	0x05	0x7F	FIR COEF A059		
6	0x06	0x00	PAGE ID		
6	0x06	0x02 to 0x07	Not used		

**Table 53. Filter Bank A Memory Map**

<span id="page-10-0"></span>



## **DEFAULT FILTER PERFORMANCE**

**The FIR filter banks have factory-programmed filter designs. They are all low-pass filters that have unity dc gain. [Table 57](#page-10-3) provides a summary of each filter design, and [Figure 32](#page-10-4) shows the frequency response characteristics. The phase delay is equal to ½ of the total number of taps.**



<span id="page-10-3"></span>

<span id="page-10-1"></span>

<span id="page-10-4"></span><span id="page-10-2"></span>**Figure 32. FIR Filter Frequency Response Curves**

# **ALARMS**

**Each sensor has an independent alarm function that provides register (see [Table 48\)](#page-7-0) contains the alarm output flags and the FNCTIO\_CTRL register (see [Table 67\)](#page-12-0) provides an option for configuring one of the digital input/output lines as an alarm indicator.**

### **ALARM USE**

**The dynamic alarm setting provides the option to compare the change in the output of each sensor over a period of 48.7 ms with that sensor's xx\_ALM\_MAGN register.**

<span id="page-11-1"></span>**Table 58. XG\_ALM\_MAGN (Page 3, Base Address = 0x28)**

<span id="page-11-5"></span><span id="page-11-4"></span><span id="page-11-3"></span><span id="page-11-2"></span>

# <span id="page-11-7"></span><span id="page-11-6"></span>**ALARM REG CONFIG**

**To use the alarm function, you need to configure the alarm setting bit [11:8] in the FNCTIO\_CTRL register, then set the value of the xx\_ALM\_MAGN register as the alarm threshold, and finally set the enable bit of the relevant alarm in the ALM\_CFGN\_x register**



<span id="page-11-0"></span>



# **ALARM EXAMPLE**

**[Table 66](#page-11-8) offers an alarm configuration example, which sets the z-axis gyroscope alarm to trip when Z\_GYRO\_OUT > 131.1°/sec (0x199B).**

<span id="page-11-8"></span>

# **SYSTEM CONTROLS**

### **GENERAL-PURPOSE INPUT/OUTPUT**

**There are four general-purpose input/output pins: DIO1, DIO2, DIO3, and DIO4. The FNCTIO\_CTRL register controls the basic function of each input/output pin, which provides a number of useful functions. Each input/output pin only supports one function at a time. In cases where a single pin has two different assignments, the enable bit for the lower priority function automatically resets to zero and is disabled. The priority is (1) data ready, (2) sync clock input, (3) alarm indicator, and (4) general purpose, where 1 identifies the highest priority and 4 indicates the lowest priority.**

**Table 67. FNCTIO\_CTRL (Page 3, Base Address = 0x06)**

<span id="page-12-0"></span>

### **DATA-READY INDICATOR**

**FNCTIO\_CTRL[3:0] provide some configuration options for using one of the DIOx lines as a data ready indicator signal, which can drive the interrupt control line of a processor. The factory default assigns DIO2 as a positive polarity, data ready signal. Use the following sequence to change this assignment to DIO1 with a negative polarity: turn to Page 3 (DIN = 0x8003) and set FNCTIO\_CTRL[3:0] = 1000 (DIN = 0x8608, then DIN = 0x8700). The timing jitter on the data ready signal is ±1.4 µs.**

# **INPUT SYNC/CLOCK CONTROL**

**FNCTIO\_CTRL[7:4] provide some configuration options for using one of the DIOx lines as an input synchronization signal for sampling inertial sensor data. For example, use the following sequence to establish DIO4 as a positive polarity,** 

**input clock pin and keep the factory default setting for the data ready function: turn to Page 3 (DIN = 0x8003) and set FNCTIO\_CTRL[7:0] = 0xFD (DIN = 0x86FD, then DIN = 0x8700). Note that this command also disables the internal sampling clock, and no data sampling occurs without the input clock signal. When selecting a clock input frequency, consider the 330 Hz sensor bandwidth because undersampling the sensors can degrade noise and stability performance.**

**Data Sheet KT-EX9-2**

### **APPLICATIONS INFORMATION**

### **MOUNTING BEST PRACTICES**



**Figure 33. Mounting Example**

<span id="page-13-0"></span>**For best performance, follow these simple rules when installing the KT-EX9-2 into a system:**

1. **Eliminate opportunity for translational force (x- and y-axis direction, per figure 6) application on the electrical connector.**

2. **Isolate mounting force to the four corners, on the portion of the package surface that surrounds the mounting holes.**

3. **Use uniform mounting forces on all four corners. The suggested torque setting is 40 inch-ounces (0.285 N-m).**

**These three rules help prevent irregular force profiles, which can warp the package and introduce bias errors in the sensors. [Figure 33](#page-13-0) provides an example that leverages washers to set the package off the mounting surface and uses 2.85 mm pass-through holes and backside washers/nuts for attachment[. Figure 34](#page-14-0) an[d Figure](#page-14-1)  [35](#page-14-1) provide some details for mounting hole and connector**



**Figure 34. Suggested PCB Layout Pattern, Connector Down**

<span id="page-14-0"></span>

**Figure 35. Suggested Layout and Mechanical Design When Using for the Mating Connector**

### <span id="page-14-1"></span>**POWER SUPPLY CONSIDERATIONS**

**The KT-EX9-2 has approximately 30 μF of capacitance across the VDD and GND pins. Whereas this capacitor bank provides a large amount of localized filtering, it also presents an opportunity for excessive charging current when the VDD voltage ramps quickly.** 

**Test with a 3.3V DC regulated power supply, and under 500mA current limiting, Figure 36 offers the spikes current in power on process, Figure 37 provides more detail on the input voltage and current behavior during the whole power on process. According the two figures, in the power on process, the max spikes current is 800mA, 200us, and the power on time is about 10ms. To ensure a reliable power on process, we recommend a guaranteed power supply capacity of at least 200mA.**



**Figure 36. Transient Current Demand, Start Up**



**Figure 37. Transient Current Demand, Peak Demand**

## **X-RAY SENSITIVITY**

**Exposure to high dose rate X-rays, such as those in production systems that inspect solder joints in electronic assemblies, can affect accelerometer bias errors. For optimal performance, avoid exposing the KT-EX9-2 to this type of inspection.**

# **ACOUSTIC NOISE SENSITIVITY**

**An inertial sensor can be placed in a chamber with a loud speaker to see whether the performance of the sensor is affected by the acoustic environment that might be encountered in a missile, helicopter, or other such mission. The shape of the chamber could be such that the sound from the loud speaker is focused onto the test article.** 

**The product adopts a technical solution of internal shock absorption design in a closed chamber, and the inertial sensors are packaged in a vacuum ceramic tube shell. These measures greatly improve the product's noise resistance performance. After an acoustic noise test with a total sound pressure of 150dB, frequency from 50Hz to 10000Hz, it has been proven that the product is immune to the acoustic noise.**

# **Data Sheet KT-EX9-2**

# **OUTLINE DIMENSIONS**



**Figure 38. 24-Lead Module with Connector Interface Dimensions shown in millimeters**